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PROJECT PROPOSAL FOR EE 507

LINEAR SYSTEM IDENTIFICATION BY USING RANDOM PROCESS

Benmei Chen

P. O. BOX 241, GONZAGA U.

SPOKANE, WA 99258

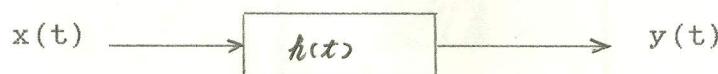
MARCH 31, 1987

## LINEAR SYSTEM IDENTIFICATION BY USING RANDOM PROCESS

### I. INTRODUCTION TO THE PROBLEM

System identification is a very important procedure in some automatic control or digital control designs. This project is to identify the unknown stable linear systems by applying the results from random process, and using the digital technique.

Consider the following system,  $h(t)$ , with the input  $x(t)$  and the output  $y(t)$ :



Let the input  $x(t)$  be a white noise. Then, using the results from random process, we can determine  $h(t)$  from  $x(t)$  and  $y(t)$ .

In order to see how this design works, we can apply it to some given systems, such as  $h(t) = u(t)$  or  $h(t) = \exp(-at) u(t)$ , where  $u(t)$  is the unit step function. And we can also compare the computed results to the exact systems.

### II. SOLUTION TO THE PROBLEM

Recall the results in EE 507 notes, the input-output cross-

correlation function of the system above is given as

$$\begin{aligned} R_{xy}(\tau) &= E [ x(t) \cdot y(t+\tau) ] \\ &= E [ x(t) \int_0^\infty x(t+\tau-a) \cdot h(a) da ] \\ &= \int_0^\infty E [ x(t) \cdot x(t+\tau-a) ] \cdot h(a) da \\ &= \int_0^\infty R_x(\tau-a) \cdot h(a) da \quad \dots \dots \quad (1) \end{aligned}$$

If the input  $x(t)$  is a white noise with the autocorrelation function

$$R_x(\tau) = Q \delta(\tau)$$

then the Equation (1) becomes

$$\begin{aligned} R_{xy}(\tau) &= \int_0^\infty Q \delta(\tau-a) \cdot h(a) da \\ &= Q h(\tau) \end{aligned}$$

For a large T, we have

$$\begin{aligned} h(\tau) &= R_{xy}(\tau) / Q \\ &\cong [ \int_0^T x(t) \cdot y(t+\tau) dt ] / (QT) \\ &= [ \int_0^T x(t-\tau) \cdot y(t) dt ] / (QT) \end{aligned}$$

Sampling both the input and output of the system by the time period of  $\Delta t$ , it is easy to see that

$$h(\tau) \cong [ \sum_{k=1}^{T/\Delta t} x(k\Delta t - \tau) \cdot y(k\Delta t) \cdot \Delta t ] / (QT) \quad \dots \quad (2)$$

Now we can write a computer program for the Equation (2) to

obtain the discrete values of  $h(t)$ . Furthermore, we can find out the z-transfer function of the system as well.

One of the key problems for this project is to generate the white noise. Recalling the results in Sections 6.1 to 6.3 in the text, we know that the process  $Z(t)$  is an idealized model ( for  $t > 0$  ) for a thermal-noise ( we can treat it as a white noise ). And  $Z(t)$  is the derivative of the Wiener process and is defined as

$$Z(t) \triangleq \lim_{\Delta t \rightarrow 0} Z_\Delta(t)$$

and

$$Z_\Delta(t) \triangleq \sum_{n=0}^{\infty} Z_n \cdot \Delta w \cdot \delta(t - n\Delta t), \quad t \geq 0,$$

$$\text{Prob}\{Z_n = +1\} = \text{Prob}\{Z_n = -1\} = 0.5$$

where  $\Delta t$  and  $\Delta w$  are known as constants and  $\delta(\cdot)$  is the impulse function.  $Z_n$  is the discrete-time  $\pm 1$  Bernoulli process.

Since the RND function ( available almost in any computer ) is uniformly distributed on  $[0, 1]$ , we can use it to generate the discrete-time Bernoulli process and then the white noise.

### III. LIMITATION

According to the sampling theorem, the maximum time interval of sampling is  $1 / (2B)$ , where  $B$  is the bandwidth of the signal. However in this particular problem we did, the bandwidth of the

white noise is infinite if it is ideal. And the bandwidth of the system  $h(t)$  is unknown. So, the computed results of Equation (2) may be unreasonable for some systems.

#### IV. ESTIMATION OF EFFORT

I think this project will require more than 40 hours of effort for

- (1) writing and debudding the program.
- (2) applying the project to some systems.
- (3) writing the project report.

#### V. COMPLETION DATE

April 16, 1987

Bernie,  
you could check the  
"whiteness" of your simulated  
"white noise" by checking its  
autocorrelation.  
Looks like a very practical  
little project. Good.  
Joe F.  
90%

\* \* \* \* \*

THANK YOU !

RECEIVED  
MAY 12 1987

\* \* \* \* \*

PROJECT REPORT FOR EE 507

LINEAR SYSTEM IDENTIFICATION BY USING RANDOM PROCESS

Benmei Chen

P.O. Box 241, GONZAGA U.

SPOKANE, WA 99258

APRIL 21, 1987

Good Beaurie  
B+ for paper  
A for course

## LINEAR SYSTEM IDENTIFICATION BY USING RANDOM PROCESS

ABSTRACT -- A method to identify the linear system by employing the sampled-hold white noise has been introduced and two particular systems have been checked.

### I. INTRODUCTION

CONSIDER the following linear time-invariant system [1] in Fig.1,

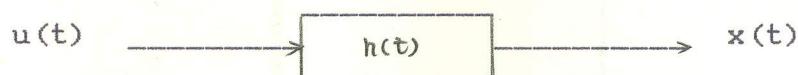


Fig. 1 The linear time-invariant system

where  $u(t)$  is the input signal,  $x(t)$  is the output and  $h(t)$  is the impulse response of the system. Assume both  $h(t)$  and  $u(t)$  are causal, it can be obtained that [2],

$$\begin{aligned} x(t) &= \int_0^{\infty} u(\tau) \cdot h(t-\tau) d\tau \\ &= \int_0^t u(\tau) \cdot h(t-\tau) d\tau \\ &= \int_0^t u(t-\tau) \cdot h(\tau) d\tau \quad \dots \end{aligned} \quad (1)$$

The impulse response of system,  $h(t)$  may be determined by applying an impulse signal to the input. However, it is difficult to have an ideal impulse signal in real world, and it is difficult to process the procedure in computer too.

This project is introducing a method to identify the system by using the random process and the computing technique.

## II. SOLUTION

### \$1. The theoretical solution:

Recall the result in notes of EE 507, the input-output cross-correlation function of the system [1] is given by

$$\begin{aligned} R_{ux}(\tau) &= E[u(t) \cdot x(t+\tau)] \\ &= E[u(t) \cdot \int_0^{\infty} u(t+\tau-a) \cdot h(a) da] \\ &= \int_0^{\infty} E[u(t) \cdot u(t+\tau-a)] \cdot h(a) da \\ &= \int_0^{\infty} R_u(\tau-a) \cdot h(a) da \quad \dots \end{aligned} \quad (2)$$

If the input  $u(t)$  is a white noise with an autocorrelation function,

$$R_u(\tau) = Q \cdot \delta(\tau)$$

where  $Q$  is a constant and  $\delta(\cdot)$  is an impulse function, then Equation (2) becomes,

$$\begin{aligned} R_{ux}(\tau) &= \int_0^{\infty} Q \cdot \delta(\tau-a) \cdot h(a) da \\ &= Q \cdot h(\tau) \quad \dots \end{aligned} \quad (3)$$

From the equation (3), it is easy to see,

$$h(\tau) = R_{ux}(\tau)/Q \quad \dots \quad (4)$$

In practice, however, we won't be able to have an ideal white noise with an impulse autocorrelation function. And it is impossible to process the ideal white noise in computer as well.

Following, we introduce a continuous-time-discrete-magnitude

white process ( see Figure 2 below ). We may call it by the sampled-hold white process because it can be obtained by sampling the ideal white noise and holding the value for a time period  $t_s$ .

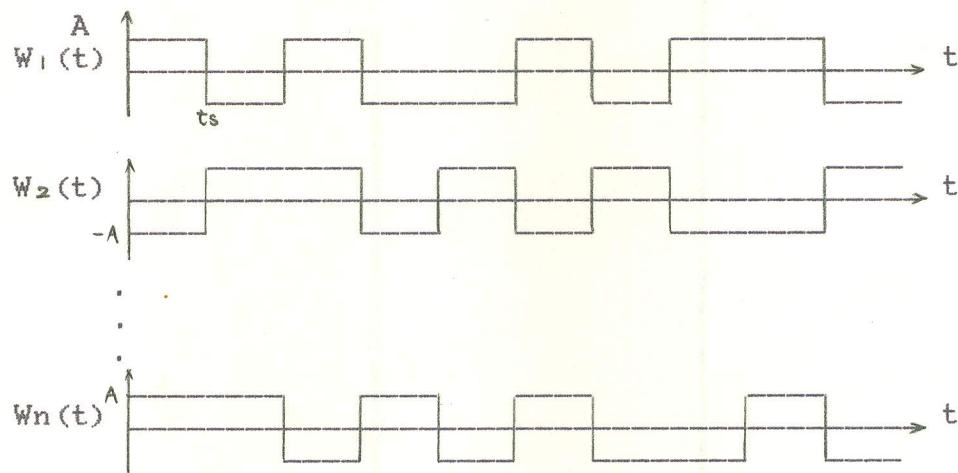


Figure 2: The sampled-hold white process

The mean and the autocorrelation function of the sampled-hold white process are known as [1],

$$E[W(t)] = 0$$

$$R_w(\tau) = \begin{cases} A^2 (\tau/t_s + 1) & -t_s < \tau \leq 0 \\ A^2 (-\tau/t_s + 1) & 0 < \tau < t_s \\ 0 & \text{elsewhere} \end{cases}$$

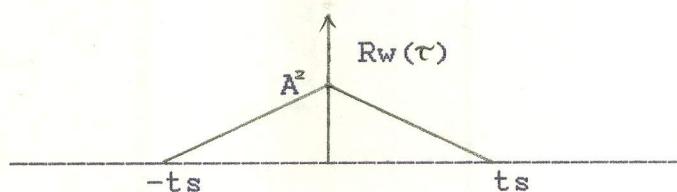


Figure 3: The autocorrelation function of sampled-hold white process

Now, we can apply the sampled-hold white process as the input of the system in Figure 1, from Equation (2), that is

$$\begin{aligned} R_{ux}(\tau) &= \int_0^\infty R_u(\tau-a) \cdot h(a) da \\ &= \int_{\tau-ts}^{\tau+ts} R_w(\tau-a) \cdot h(a) da \quad \dots \end{aligned} \quad (5)$$

If  $ts$  is small enough and  $h(a)$  is a continuous function, then  $h(\tau)$  can be regarded as a constant  $h_\tau$  in the interval  $[\tau-ts, \tau+ts]$ . So that the equation (5) can be written as,

$$\begin{aligned} R_{ux}(\tau) &\cong h_\tau \cdot \int_{\tau-ts}^{\tau+ts} R_w(\tau-a) da \\ &\cong h_\tau \cdot ts \cdot A^2 \end{aligned} \quad \dots \quad (6)$$

From Eq. (6)

$$h_\tau \cong R_{ux}(\tau) / (A^2 \cdot ts) \quad \dots \quad (7)$$

Assume the  $u(t)$  and  $x(t)$  are ergodic, then for a large  $T$ ,

$$\begin{aligned} h_\tau &\cong [\int_0^T u(t) \cdot x(t+\tau) dt] / (A^2 \cdot ts \cdot T) \\ &= [\int_0^T u(t-\tau) \cdot x(t) dt] / (A^2 \cdot ts \cdot T) \quad \dots \end{aligned} \quad (8)$$

The integration in Equation (8) can be approximated by

$$h_\tau \cong [\sum_{k=0}^{T/\Delta t} u(k \cdot \Delta t - \tau) \cdot x(k \cdot \Delta t) \cdot \Delta t] / (A^2 \cdot ts \cdot T)$$

Let  $\Delta t = ts$ , we have

$$h_\tau \cong [\sum_{k=0}^{T/\Delta t} u(k \cdot ts - \tau) \cdot x(k \cdot ts)] / (A^2 \cdot T) \quad \dots \quad (9)$$

The Equation (9) is easy to be processed in computer.

## \$2. The practical solution:

Here, two computing algorithms are presented. One is to generate the sampled-hold white process  $W(t)$ , and the other is to identify the system impulse response  $h(t)$ .

(1) An algorithm for creating sampled-hold white process  $W(t)$  with sampling period  $ts$  and time internal  $[0, T]$ ,

```
For t=0 sec. to T sec. step ts sec. Do
  If RND<.5 then
    W=A ( a constant, the magnitude of W(t) )
  else
    W=-A
  End if
  W(t)=W
End of i loop
```

In this algorithm, the RND is a random variable uniformly distributed on  $[0, 1]$ . It is almost available in all computers.

(2) The algorithm to obtain the discrete system impulse response  $h(\tau)$  for a sampling period  $ta$  and  $0 < t < T_0$ .

```
For τ=0 sec. to T0 sec. step ta sec. Do
  h(τ) = 0 ( initial )
  For t=τ sec. to T sec. step ts sec. Do
    h(τ) = h(τ) + u(t) * x(t-τ)
```

End of t loop

$$h(\tau) = h(\tau) / (A^2 * T)$$

End of  $\tau$  loop

This algorithm is corresponding to the Equation (9). The general computer program in VAX-11 FORTRAN is given in Appendix A.

### III. RESULTS

In order to check how this method works, we applied it to the following two particular systems:

$$H_a(s) = (2s + .8) / (s + .8s + .5) \dots \quad (S1)$$

and

$$H_c(s) = (2s - .8) / (s - .8s + 1.16) \dots \quad (S2)$$

Take the inverse Laplace transform of both two systems, we have the impulse system responses,

$$h_a(t) = \exp(-.3t) + \exp(-.5t) \dots \quad (H1)$$

and

$$h_c(t) = 2 \exp(-.4t) \cos(t) \dots \quad (H2)$$

Recall the Equation (1) in part I,

$$x(t) = \int_0^t u(\tau) \cdot h(t-\tau) \cdot d\tau$$

Let  $t = n \cdot ts$ ,

$$\begin{aligned}
x(t) &= \int_0^{ts} u(\tau) \cdot h(t-\tau) d\tau + \int_{ts}^{zts} u(\tau) \cdot h(t-\tau) d\tau \\
&\quad + \dots + \int_{(n-1)ts}^{nts} u(\tau) \cdot h(t-\tau) d\tau \\
&= u_0 \cdot \int_0^{ts} h(t-\tau) d\tau + u_1 \cdot \int_{ts}^{zts} h(t-\tau) d\tau \\
&\quad + \dots + u_{n-1} \cdot \int_{(n-1)ts}^{nts} h(t-\tau) d\tau \\
&= u_0 \cdot [H(ts) - H((n-1)ts)] + u_1 \cdot [H((n-1)ts) - H((n-2)ts)] \\
&\quad + \dots + u_{n-1} \cdot [H(ts) - H(0)] \quad \dots \quad (10)
\end{aligned}$$

where  $u_k$ ,  $k=0, 1, \dots, n-1$ , is hold to be a constant within the sampled period and  $H(\cdot)$  is the original function of  $h(t)$ ,  $H'(t) = h(t)$ .

From the Equation (10), we can obtain the exact output of the systems if we know the original function of  $h(t)$ .

Apply this and the algorithms we had in part II to the systems (S1) and (S2), we got very reasonable results ( see Fig. 4-7 and the data in Appendix B ).

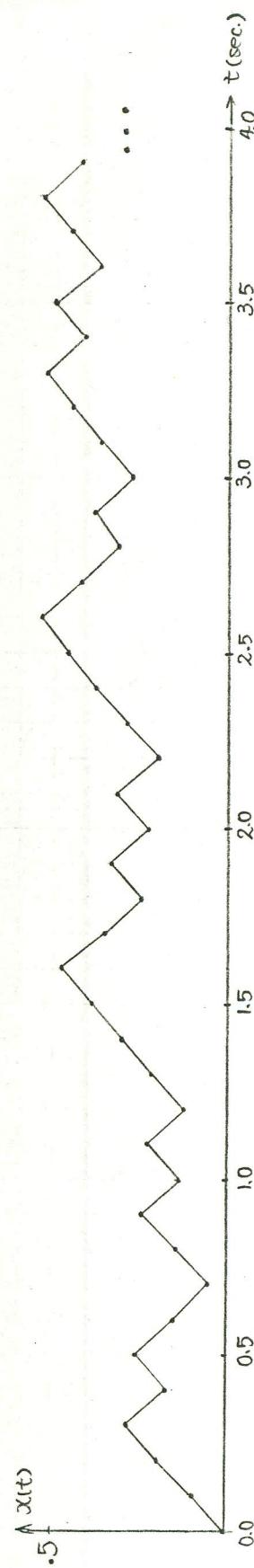
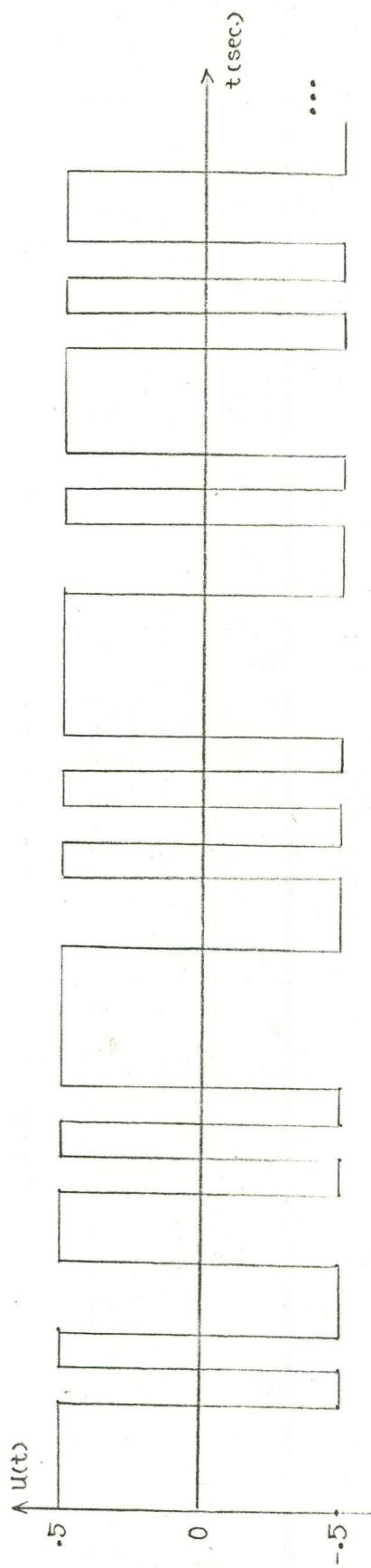


FIGURE 4: The input and output signals for system (S1)

$$H_a(s) = (2s + .8) / (s^2 + .8s + .5)$$

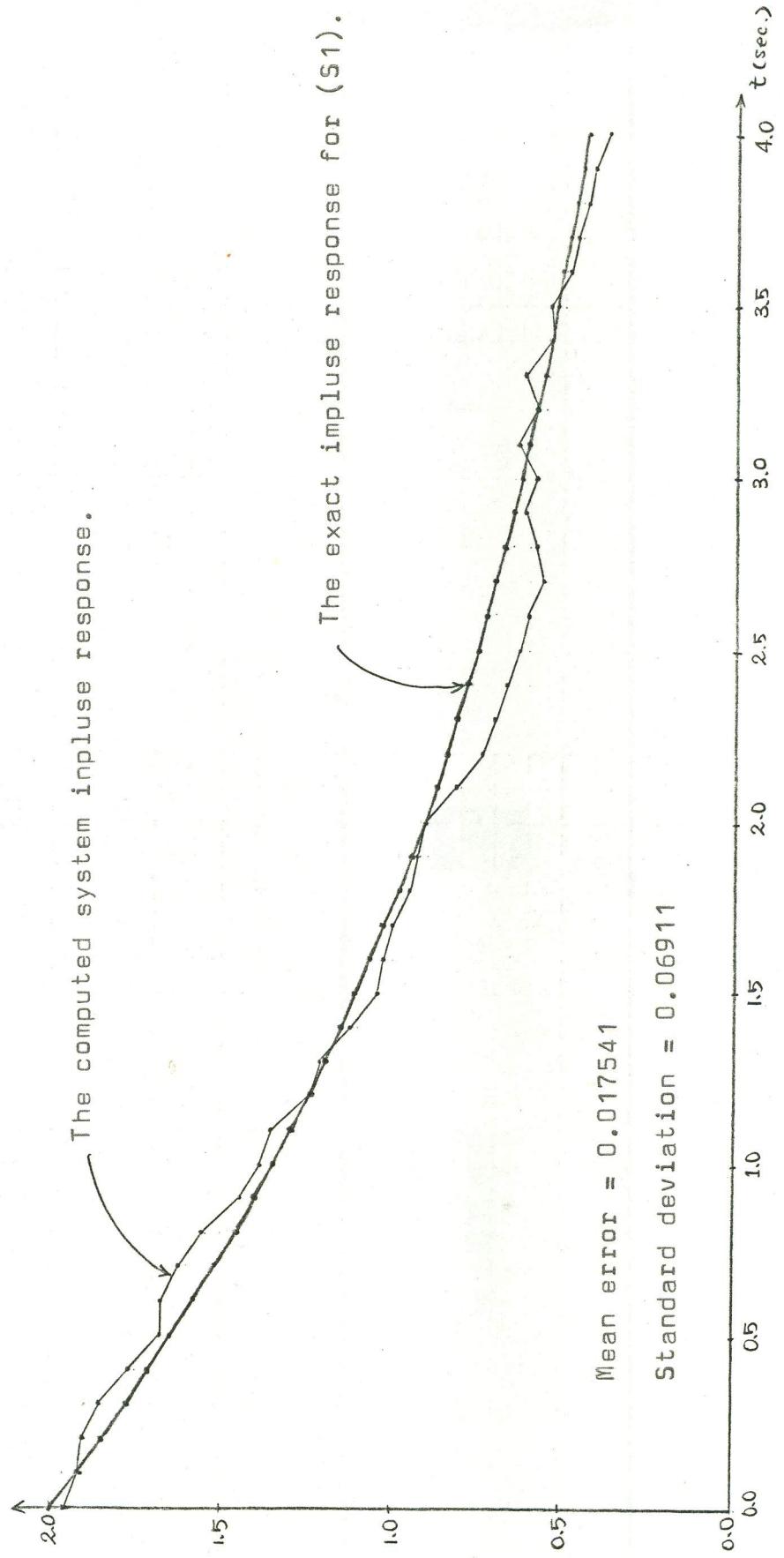


Figure 5: The exact and computed impulse responses for system (S1).

$$H_a(s) = (s + 0.8) / (s^2 + 0.8s + 5)$$

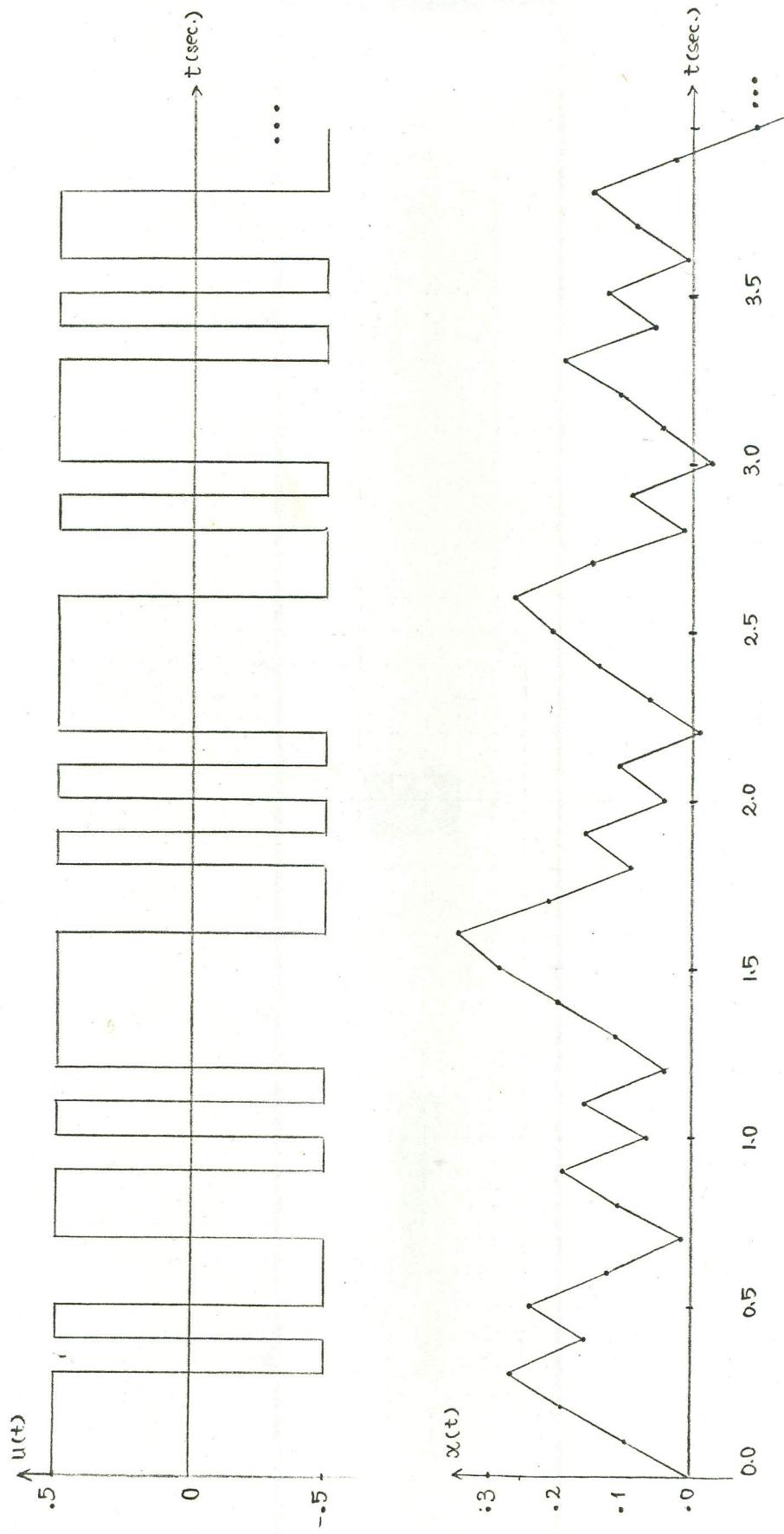


Figure 6: The input and output signals for (S2)

$$H_C(s) = (2s - .8) / (s^2 - .8s + 1.16)$$

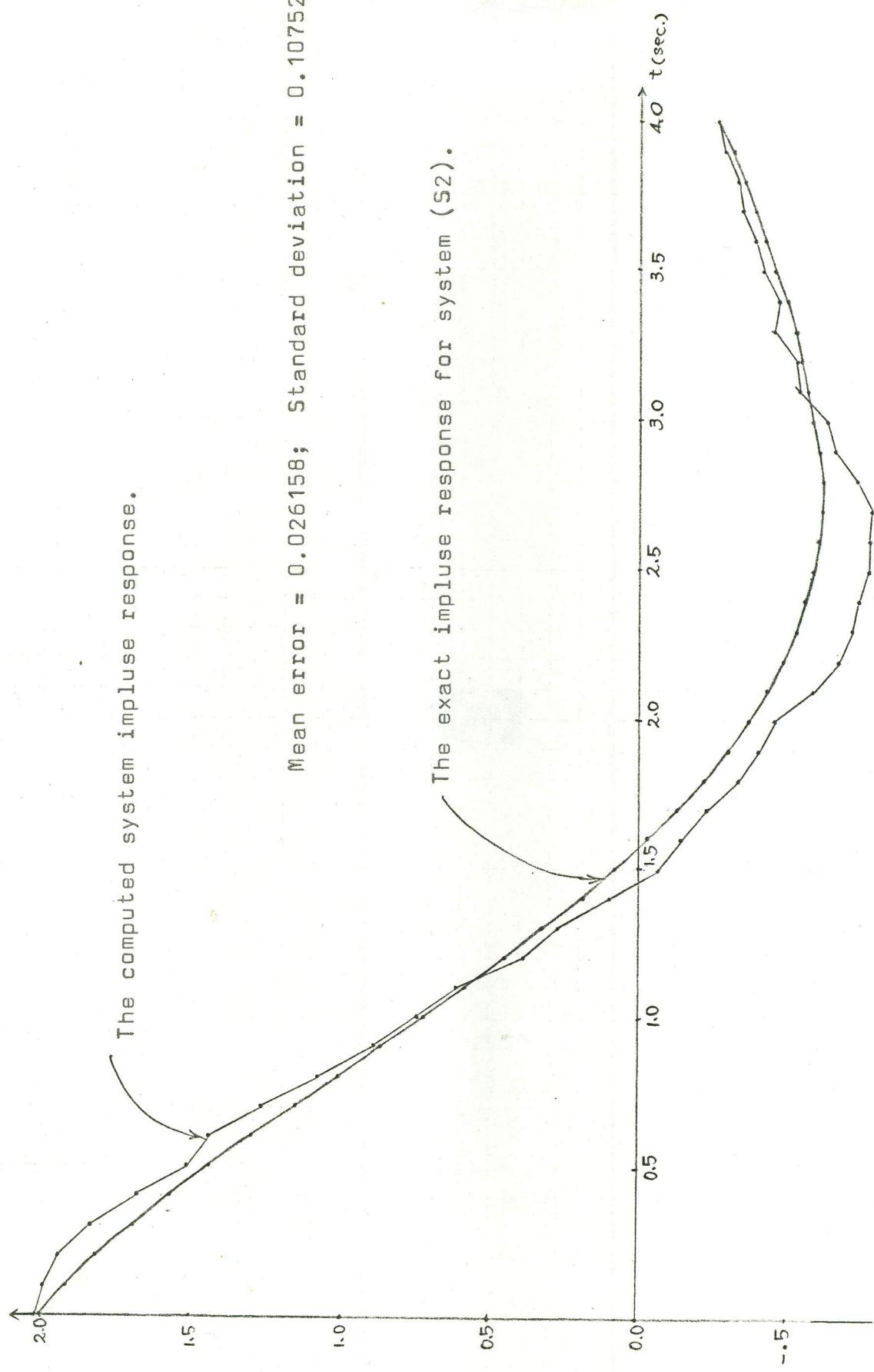


Figure 7: The exact and computed impulses responses for system (S2)

## VI. APPLICATIONS

The main application of this project is to identify the systems. However, we can use this project to construct a transversal filter corresponding to the interested systems too. Recall the Equation (1) in part I again,

$$x(t) = \int_0^t u(t-\tau) \cdot h(\tau) d\tau$$

The equation above can be also approximated by

$$x(t) = \sum_{k=0}^{t/t_s} u(t-k \cdot t_s) \cdot h(k \cdot t_s) \cdot t_s$$

This result can be realized by using a delay line with taps at the delays  $k \cdot t_s$ ; the output of each tap is multiplied by the preset weight  $h(k \cdot t_s) \cdot t_s$ , where  $h(k \cdot t_s)$  is obtained from the system identification and  $t_s$  is given. Such a system is shown in Figure 8.

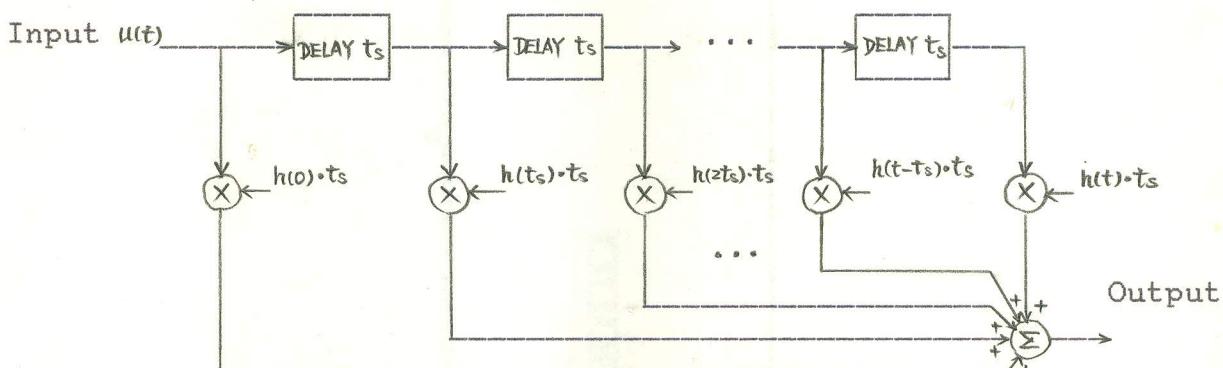


Figure 8: A tapped-delay-line (transversal) filter

A filter constructed using a tapped delay line, tap weights, and adder in the configuration shown in Figure 8 is called a transversal filter.

## V. CONCLUSIONS

In this project the sampled-hold white process had been defined and applied to identify the linear systems. Two systems had been checked turning out with nice results. And a application to construct the transversal filter had been introduced as well.

## REFERENCES

- [1] Dr. Feeley, " Notes of EE 507 "; University of Idaho, 1987
- [2] Ferrel G. Stremler, " Introduction to Communication Systems";  
New York, 1981
- [3] William A. Gardner, " Introduction to Random Processes ";  
1986
- [4] Arthur Gelb, " Applied Optimal Estimation "; 1974

APPENDIX A: THE FORTRAN PROGRAM IN VAX-11 TO IDENTIFY SYSTEMS.

```
C ****  
C * THE LINEAR SYSTEM IDENTIFICATION *  
C ****  
  
DIMENSION H(40), U(2500), X(2500)  
DO 10 I=55555555, 80000000, 10000  
Y = RAN(K)  
IF ( Y .LT. .5 ) THEN  
    Z=.5  
ELSE  
    Z=-.5  
END IF  
M = ( I-55555555)/10000  
U(M) = Z  
10 CONTINUE  
X(0) = 0  
DO 30 I=1, 2500  
    X(I) = 0  
    M = I  
    DO 20 J=1, M  
        F1 = -EXP(-.03*J)/.3 - EXP(-.05*J)/.5  
        F2 = -EXP(-.03*(J-1))/.3 - EXP(-.05*(J-1))/.5  
        X(I) = X(I) + U(I-J)*(F1-F2)  
    END DO  
30 CONTINUE  
C PRINT OUT THE INPUT AND OUTPUT SIGNALS.  
40 FORMAT ( 3F16.8 )  
DO 50 I=0, 40  
    PRINT 100, .1*I, U(I), X(I)  
50 CONTINUE  
DO 70 I=1, 40  
    H(I) = 0  
    M = I+1  
    DO 60 K=M, 2500  
        H(I)=H(I)+U(K-I-1)*X(K)  
    END DO  
60 H(I) = H(I)/(.25*250)  
70 CONTINUE  
80 FORMAT ( 4F16.8 )  
DO 90 I=1,40  
    A=EXP(-.03*I)+EXP(-.05*I)  
    B=A-H(I)  
    P=P+B  
    Q=Q+B*B  
    T=.1*I  
    PRINT 80, T, A, H(I), B  
90 CONTINUE  
PRINT 100, P/40, Q/40  
100 FORMAT ( 2F16.8 )  
END
```

## APPENDIX B: THE DATA FOR SYSTEMS (S1) &amp; (S2)

## (1) The input and output for system (S1) &amp; (S2)

TIME ( sec. )	INPUT $u(t)$	(S1) OUTPUT	(S2) OUTPUT
0.10	0.500000	0.000000	0.000000
0.20	0.500000	0.098028	0.095008
0.30	0.500000	0.192222	0.180123
0.40	-.500000	0.282740	0.255598
0.50	0.500000	0.173679	0.131799
0.60	-.500000	0.264965	0.209023
0.70	-.500000	0.156639	0.087498
0.80	0.500000	0.052571	-.022629
0.90	0.500000	0.148641	0.068583
1.00	-.500000	0.240955	0.151068
1.10	0.500000	0.133602	0.034900
1.20	-.500000	0.226540	0.120126
1.30	0.500000	0.119786	0.006761
1.40	0.500000	0.213279	0.094716
1.50	0.500000	0.303119	0.174197
1.60	0.500000	0.389460	0.245106
1.70	-.500000	0.472444	0.307773
1.80	-.500000	0.356156	0.172543
1.90	0.500000	0.244450	0.049656
2.00	-.500000	0.333194	0.129113
2.10	0.500000	0.222421	0.010863
2.20	-.500000	0.312058	0.094819
2.30	0.500000	0.202139	-.019127
2.40	0.500000	0.292595	0.068896
2.50	0.500000	0.379516	0.148764
2.60	0.500000	0.463052	0.220519
2.70	-.500000	0.543340	0.284310
2.80	-.500000	0.424461	0.150432
2.90	0.500000	0.310265	0.029051
3.00	-.500000	0.396617	0.110114
3.10	0.500000	0.283543	-.006477
3.20	0.500000	0.370968	0.079151
3.30	0.500000	0.454979	0.156874
3.40	-.500000	0.396617	0.226717
3.50	0.500000	0.417267	0.098829
3.60	-.500000	0.499535	0.173303
3.70	0.500000	0.382543	0.050173
3.80	0.500000	0.466208	0.129422
3.90	-.500000	0.546611	0.201012
4.00	-.500000	0.427833	0.075020

## APPENDIX B: THE DATA FOR SYSTEMS (S1) &amp; (S2)

(2) The Exact and computed impulse response for system (S1)

TIME ( sec. )	EXACT h(t)	COMPUTED h(t)	ERROR
0.00	2.000000	1.959443	0.040557
0.10	1.921675	1.907846	0.013829
0.20	1.846602	1.899722	-.053120
0.30	1.774639	1.847401	-.072762
0.40	1.705651	1.770145	-.064494
0.50	1.639509	1.680864	-.041355
0.60	1.576088	1.686978	-.110890
0.70	1.515272	1.621921	-.106649
0.80	1.456948	1.544329	-.087381
0.90	1.401008	1.441172	-.040164
1.00	1.347349	1.394400	-.047051
1.10	1.295874	1.361603	-.065730
1.20	1.246488	1.241553	0.004935
1.30	1.199103	1.214590	-.015487
1.40	1.153632	1.119249	0.034383
1.50	1.109995	1.037584	0.072411
1.60	1.068112	1.033676	0.034436
1.70	1.027911	1.005929	0.021981
1.80	0.989318	0.959700	0.029618
1.90	0.952267	0.934910	0.017357
2.00	0.916691	0.904056	0.012635
2.10	0.882530	0.806194	0.076336
2.20	0.849722	0.736199	0.113523
2.30	0.818213	0.699530	0.118683
2.40	0.787946	0.670090	0.117856
2.50	0.758871	0.620324	0.138547
2.60	0.730938	0.597112	0.133826
2.70	0.704098	0.558327	0.145771
2.80	0.678308	0.576432	0.101876
2.90	0.653522	0.611852	0.041670
3.00	0.629700	0.581083	0.048617
3.10	0.606802	0.633850	-.027048
3.20	0.584789	0.590923	-.006134
3.30	0.563627	0.608492	-.044865
3.40	0.543278	0.528905	0.014373
3.50	0.523712	0.533499	-.009787
3.60	0.504894	0.492902	0.011992
3.70	0.486796	0.468780	0.018016
3.80	0.469388	0.429416	0.039972
3.90	0.452641	0.409610	0.043031
4.00	0.436530	0.370649	0.065881

mean error = 0.017541

standard deviation = 6.911241E-02

## APPENDIX B: THE DATA FOR SYSTEMS (S1) &amp; (S2)

(3) The Exact and computed impulse response for system (S2)

TIME ( sec. )	EXACT h(t)	COMPUTED h(t)	ERROR
0.00	2.000000	2.067218	-.067218
0.10	1.911979	1.994654	-.082675
0.20	1.809431	1.944903	-.135472
0.30	1.694615	1.835174	-.140559
0.40	1.569753	1.689932	-.120179
0.50	1.437008	1.518912	-.081904
0.60	1.298464	1.434804	-.136340
0.70	1.156111	1.275768	-.119657
0.80	1.011826	1.096838	-.085012
0.90	0.867365	0.892931	-.025566
1.00	0.724351	0.742459	-.018108
1.10	0.584265	0.610401	-.026136
1.20	0.448442	0.394151	0.054291
1.30	0.318067	0.272902	0.045165
1.40	0.194174	0.092088	0.102086
1.50	0.077643	-.070643	0.148286
1.60	-.030793	-.147355	0.116561
1.70	-.130550	-.236387	0.105837
1.80	-.221182	-.334957	0.113775
1.90	-.302383	-.402862	0.100478
2.00	-.373974	-.467548	0.093574
2.10	-.435895	-.589816	0.153921
2.20	-.488201	-.677538	0.189338
2.30	-.531047	-.722059	0.191011
2.40	-.564686	-.746970	0.182284
2.50	-.589449	-.785974	0.196526
2.60	-.605743	-.786904	0.181162
2.70	-.614038	-.795346	0.181308
2.80	-.614856	-.742403	0.127546
2.90	-.608764	-.665871	0.057107
3.00	-.596360	-.647548	0.051188
3.10	-.578268	-.541427	-.036841
3.20	-.555126	-.530746	-.024380
3.30	-.527581	-.457152	-.070429
3.40	-.496278	-.479008	-.017271
3.50	-.461855	-.412421	-.049434
3.60	-.424934	-.393095	-.031839
3.70	-.386119	-.356956	-.029163
3.80	-.345988	-.333277	-.012711
3.90	-.305089	-.290464	-.014625
4.00	-.263937	-.270497	0.006560

mean error = 0.026158

standard deviation = .1075275