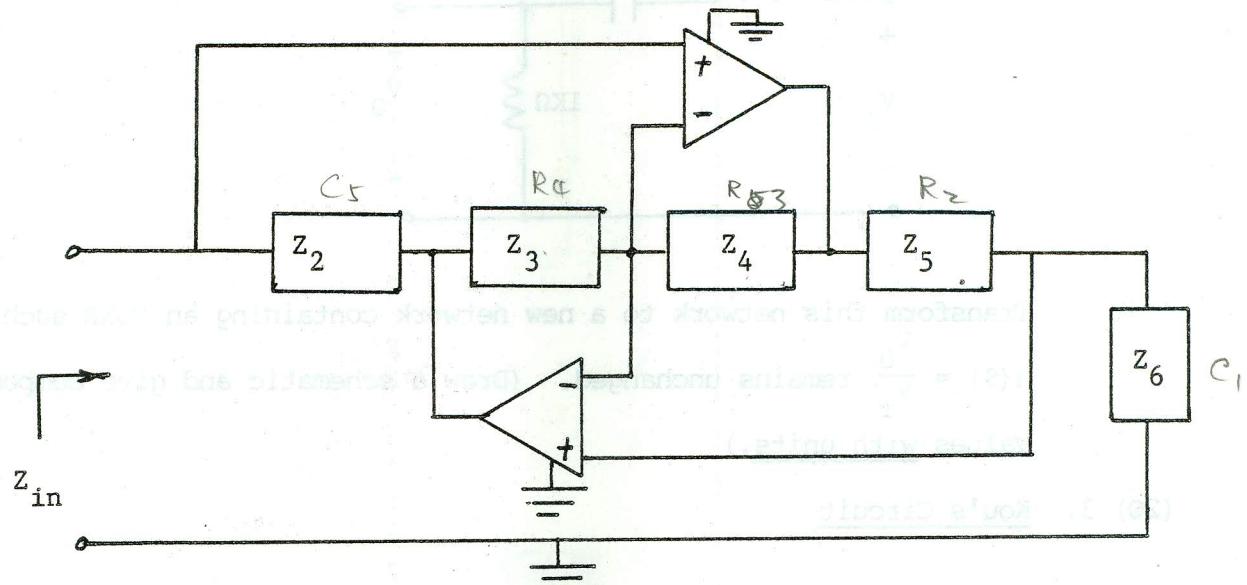


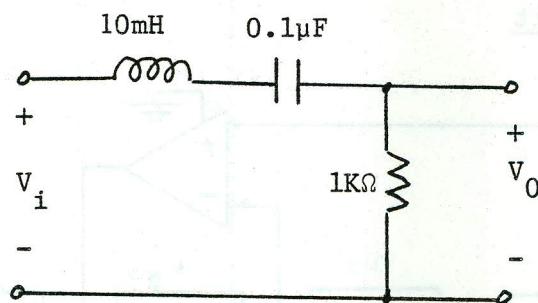
points(25) 1. Antoniou's GIC Circuit

- $Z_6$  is the impedance of a 1K ohm resistor. Find component values (not impedances) such that  $Z_{in}$  is the impedance of a  $1\mu F$  capacitor.
- $Z_6$  is the impedance of a  $1\mu F$  capacitor. Find component values (not impedances) such that  $Z_{in}$  is the impedance of a  $1nF^2$  FDNR.
- Find component values to simulate a  $1mH$  inductor.

Page 2

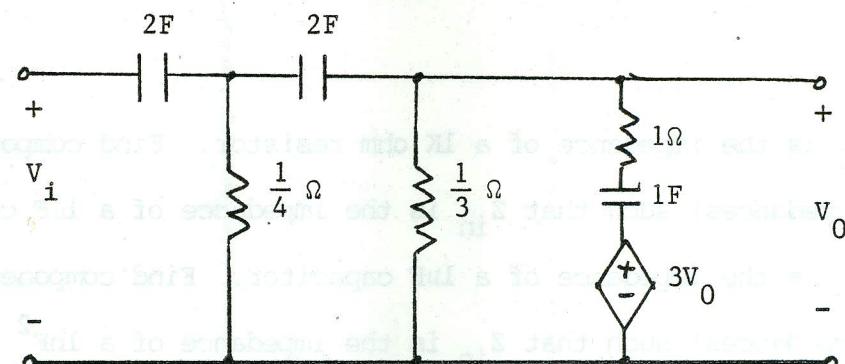
(points)

(20) 2. Frequency-dependent Negative Resistor



Transform this network to a new network containing an FDNR such that  
 $H(s) = \frac{V_0}{V_i}$  remains unchanged. (Draw a schematic and give component values with units.)

(20) 3. Kou's Circuit



For the above network

$$H(s) = \frac{V_0}{V_i} = \frac{s^2}{s^2 + 3s + 3}$$

Perform appropriate frequency and impedance scaling to realize the following transfer function using practical component values.

$$H(s) = \frac{s^2}{s^2 + 3 \times 10^{-3} s + 3 \times 10^{-6}}$$

(Draw a schematic with component values.)

Points

(20) 4. RC Decomposition

$$Z(s) = \frac{s - 1}{(s + 1)(s + 2)}$$

Find the RC decomposition (i.e., find realizations for  $Z_{RC}^{(i)}$  and  $Z_{RC}^{(ii)}$ )

(20) 5.  $\omega_0$  and Q Sensitivities

$$H(s) = \frac{(R+1)}{(R_1+1)} \left[ \frac{\frac{s^2}{R_1(R+1)}}{s^2 + \frac{R_1(R+1)}{(R_1+1)} s + R} \right]$$

$$\text{Find } S_R^{\omega_0}, S_{R_1}^{\omega_0}, S_R^Q \text{ and } S_{R_1}^Q$$

(20) 6. Multiple-amplifier Biquad

Suppose

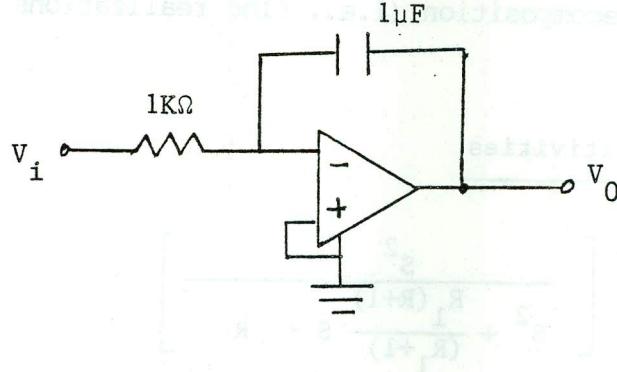
$$\frac{V_1}{V_i} = \frac{-k_1 s}{s^2 + b_1 s + b_0}$$

$$\frac{V_3}{V_i} = \frac{-k_2}{s^2 + b_1 s + b_0}$$

Show how  $V_i$ ,  $V_1$  and  $V_3$  can be summed to realize the transfer function of a HPF. Find  $k_1$  and  $k_2$  in terms of  $b_0$  and  $b_1$ .

**Page 4**  
**Points**

(25) 7. Non-ideal Integrator



$$A(s) = \frac{10^7}{s + 100}$$

The transfer function of the non-ideal integrator is

$$H(s) = -\frac{1}{RC} \frac{A_0 \omega_a}{s^2 + A_0 \omega_a s + \frac{\omega_a}{RC}}$$

- determine the pole locations
- suppose we have a magnitude-response tolerance of  $\epsilon_p = 2\%$  and a phase response tolerance of  $\epsilon_\phi = 3^\circ$ . What is the useful frequency range of the integrator?

$$1. (a) Z_6 = R_6 = 1000\Omega$$

$$\text{LET } Z_2 = R_2 = Z_3 = R_3 = Z_5 = R_5 = 1000\Omega \quad \checkmark$$

$$Z_4 = \frac{1}{C_4 S}, \quad C = \frac{C_4 \cdot R_3 \cdot R_5}{R_2 \cdot R_L}$$

$$C_4 = \frac{C R_2 R_L}{R_3 \cdot R_5} = C = 1 \mu F \quad \checkmark$$

$$(b) D = \frac{R_2 R_4 C_1 C_5}{R_3} = 1 nF^2$$

$$\text{LET } R_3 = R_4 = R_5 = 1000\Omega \quad C_2 = 1 \mu F \quad \checkmark$$

$$C_6 = \frac{D R_4}{R_5 R_3 C_2} = \frac{1 \times 10^{-9} \times 10^3}{10^3 \times 10^3 \times 10^{-6}} = 1 \mu F$$

$$(c) \frac{R_2 R_4 R_6 C_5}{R_3} = 1 mH$$

$$\text{LET } R_2 = R_4 = R_6 = R_3 = 1 k\Omega \quad \checkmark$$

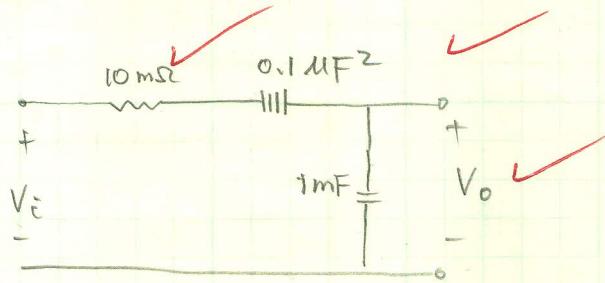
$$C_5 = \frac{10^3 \times 10^{-3}}{10^3 \times 10^3 \times 10^3} = 1 nF \quad \checkmark$$

\* FOR THE RESISTORS  $Z_x = R_x$

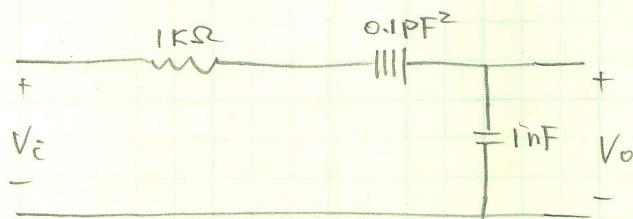
FOR CAPACITORS  $Z_x = \frac{1}{C_x S}$

$x = 1, 2, 3, 4, 5, 6$

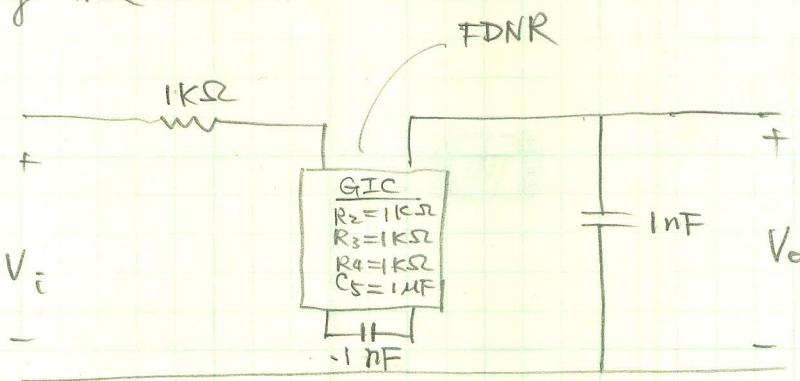
2. Divided each element in circuit by  $s$ , we have



Impedance scale by  $10^6$ , we have



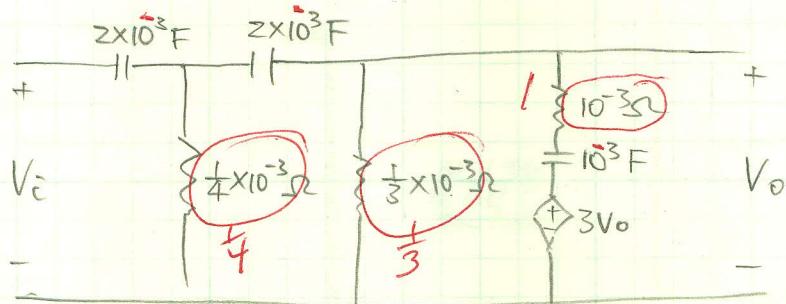
using the FDNR



3.

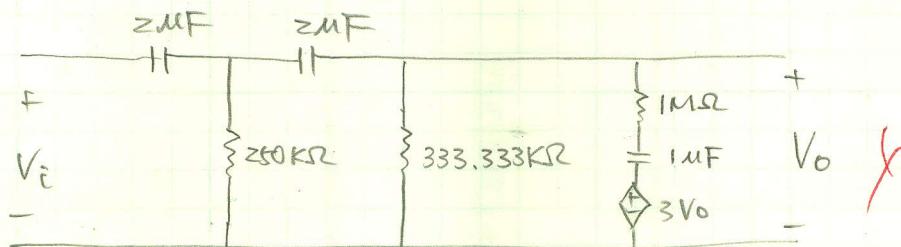
$$\begin{aligned}
 H(s) &= \frac{s^2}{s^2 + 3 \times 10^{-3}s + 3 \times 10^{-6}} \\
 &= \frac{s^2 \times 10^6}{s^2 \times 10^6 + 3 \times 10^3 s + 3} \\
 &= \frac{(1000s)^2}{(1000s)^2 + 3(1000s) + 3} \\
 &= \frac{s^2}{s^2 + 3s + 3} \quad s = 1000s
 \end{aligned}$$

So, we need to scale the frequency in original circuit by a factor  $10^{-3}$



- 15

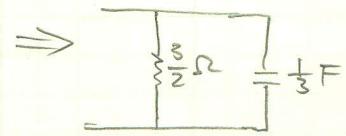
Impedance scale by  $10^9$



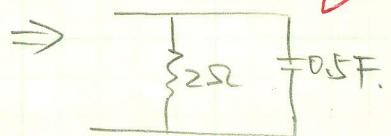
4.

$$\begin{aligned} Z(s) &= \frac{s-1}{(s+1)(s+2)} \\ &= \frac{3}{s+2} - \frac{2}{s+1} \\ &\triangleq Z_{RC}^{(i)} - Z_{RC}^{(ii)} \end{aligned}$$

$$Z_{RC}^{(i)} = \frac{3}{s+2} = \frac{1}{\frac{2}{3}s + \frac{1}{3}}$$



$$Z_{RC}^{(ii)} = \frac{2}{s+1} = \frac{1}{0.5 + 0.5s}$$



5.  $\omega_0 = \sqrt{\frac{1}{LC}} = \sqrt{\frac{1}{R^2 C}}$

$$\frac{\partial \omega_0}{\partial R} = \frac{1}{2} \cdot \frac{1}{\sqrt{R^2 C}} = \frac{1}{2} \cdot \frac{1}{\sqrt{R}} = \frac{\omega_0}{2R}$$

$$S_R \stackrel{\omega_0}{=} \frac{R}{\omega_0} \cdot \frac{\partial \omega_0}{\partial R} = \frac{R}{\omega_0} \cdot \frac{\omega_0}{2R} = 0.5$$

$$S_{R_1} \stackrel{\omega_0}{=} \frac{R_1}{\omega_0} \cdot \frac{\partial \omega_0}{\partial R_1} = 0$$

$$Q = \sqrt{\frac{R_1 + 1}{R_1(R + 1)}}$$

$$\frac{\partial Q}{\partial R} = \frac{1}{2} \cdot \frac{1}{R} \sqrt{\frac{R_1 + 1}{R_1(R + 1)}} - \sqrt{\frac{R_1 + 1}{R_1(R + 1)}} \cdot \frac{1}{R(R + 1)^2}$$

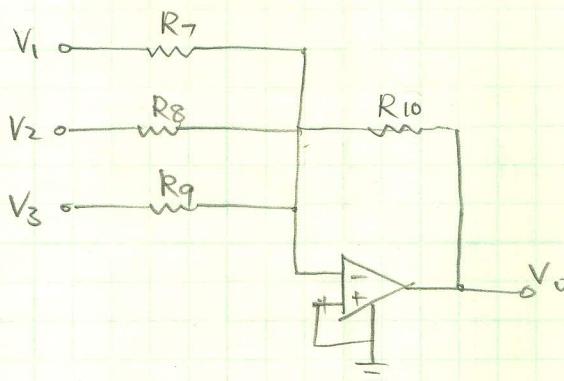
$$= \frac{1}{2} \cdot \frac{1}{R} \cdot Q - Q \cdot \frac{1}{R + 1} = -\frac{1 \cdot (R - 1)}{2R(R + 1)} \cdot Q$$

$$S_R \stackrel{Q}{=} \frac{\partial R}{Q} - \frac{\partial Q}{\partial R} = -\frac{1}{2} \cdot \frac{1 \cdot (R - 1)}{R + 1} \approx -\frac{1}{2} \quad \text{IF } R \gg 1$$

$$\frac{\partial Q}{\partial R_1} = \frac{\sqrt{\frac{R_1 + 1}{R_1}}}{R + 1} \cdot \frac{\partial \frac{R_1 + 1}{R_1}}{\partial R_1} = -\frac{\sqrt{\frac{R_1 + 1}{R_1}}}{R + 1} \cdot \frac{1}{R_1^2} = -Q \cdot \frac{1}{R_1(R_1 + 1)}$$

$$S_{R_1} \stackrel{Q_1}{=} -\frac{R_1}{Q} \cdot \frac{\partial Q}{\partial R_1} = \frac{R_1}{Q} \cdot (-Q \cdot \frac{1}{R_1(R_1 + 1)}) = -\frac{1}{R_1 + 1} = 0 \quad \text{IF } R_1 \gg 1$$

6. Using the summer below.



the output of the summer is

$$\begin{aligned}
 \frac{V_0}{V_i} &= -\frac{R_{10}}{R_7} \frac{V_1}{V_i} - \frac{R_{10}}{R_8} \frac{V_2}{V_i} - \frac{R_{10}}{R_9} \frac{V_3}{V_i} \\
 &= -\frac{R_{10}}{R_7} \cdot \frac{-k_1 s}{s^2 + b_1 s + b_0} - \frac{R_{10}}{R_8} \cdot \frac{-k_2}{s^2 + b_1 s + b_0} - \frac{R_{10}}{R_9} \\
 &= -\frac{\frac{R_{10}}{R_9} s^2 + (\frac{R_{10}}{R_9} b_1 - \frac{R_{10} k_1}{R_7})s + (\frac{R_{10}}{R_9} b_0 - \frac{R_{10} k_2}{R_8})}{s^2 + b_1 s + b_0}
 \end{aligned}$$

IF WE PICK  $R_7 = R_8 = R_9 = 1000\Omega$ , we have

$$\frac{V_0}{V_i} = -\frac{\frac{R_{10}}{1000} s^2 + (b_1 - k_1)s + (b_0 - k_2)}{s^2 + b_1 s + b_0} \quad \checkmark$$

IN ORDER TO OBTAIN A HPF,

WE have to let  $k_1 = b_1$ ,  $k_2 = b_0$  ✓

However. in general,

$$k_1 = \frac{R_7}{R_9} b_1$$

$$k_2 = \frac{R_8}{R_9} \cdot b_0$$

where  $R_8, R_9$  can be any reasonable values.

7. From givens, we have.

$$A_0 \omega_a = 10^7 \quad \omega_a = 100$$

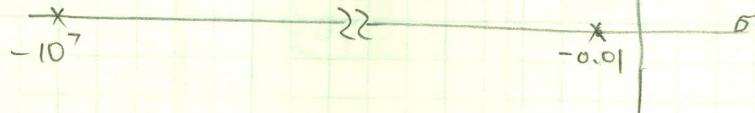
$$\text{so, } A_0 = 10^7 / 100 = 10^5$$

a) The pole locations

$$\begin{aligned} s_{1,2} &= -\frac{A_0 \omega_a}{2} \pm \frac{1}{2} \sqrt{A_0^2 \omega_a^2 - \frac{4 \omega_a}{RC}} \\ &= -\frac{10^5 \times 100}{2} \pm \frac{1}{2} \sqrt{(10^7)^2 - \frac{4 \times 100}{10^3 \times 10^{-6}}} \\ &= -5 \times 10^6 \pm 4999999.99 \end{aligned}$$

$$s_1 = -9999999.99 \approx -10^7 \quad \checkmark$$

$$s_2 = -0.01 \quad \checkmark$$



$$\begin{aligned} (b) \quad \omega_{m1} &= \frac{1}{RCA_0} \cdot \frac{1}{\sqrt{0.02 \epsilon_p}} \\ &= \frac{1}{10^3 \times 10^{-6} \times 10^5} \cdot \frac{1}{\sqrt{0.02 \times 2}} = 0.05 \text{ rad/sec.} \end{aligned}$$

$$\omega_{m2} = \frac{1}{A_0 RC} \tan(90^\circ - 3^\circ) = 0.1908 \text{ rad/sec.}$$

$$\omega_{M1} = A_0 \omega_a \sqrt{0.02 \epsilon_p} = 10^5 \times 100 \sqrt{0.02 \times 2} = 2 \times 10^6 \text{ rad/sec.}$$

$$\omega_{M2} = A_0 \omega_a \tan 3^\circ = 524077.7928 \text{ rad/sec.}$$

So, useful freq. range

$$1908 \text{ rad/sec.} \rightarrow 524077.7928 \text{ rad/sec.}$$